

RL-based Resource Efficient Actuation in UAV-aided Sensor-Actuator Networks

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Abstract—Integrating unmanned aerial vehicles (UAVs) into wireless sensor-actuator networks (WSANs) offers flexibility and improved system performance. However, imprecise localization and limited battery capacity constrain UAV operations. This paper explores the use of battery swap station (BSS)-assisted UAV to facilitate timely actuation in WSANs while addressing the above limitations. The UAV collects data via backscatter communication from the sensor nodes and delivers to the energy-constrained actuator nodes along with the required energy. Incorporating UAV location uncertainty and Nakagami- m wireless channel fading, closed-form expressions are derived for the ergodic capacity in backscatter communication and the expected energy harvesting rate. To minimize the maximum delay in actuation, an optimization problem is formulated. To reduce complexity, the problem is transformed into an equivalent node visit sequence optimization and solved using sequential deep reinforcement learning (SDRL). Through Monte Carlo simulations, the accuracy of our analysis is verified. Our results confirm that the proposed SDRL based strategy consistently offers reduced actuation delay with a significantly less computation overhead.

Index Terms—UAV, wireless sensor-actuator network, backscatter communication, wireless energy transfer, DRL

I. INTRODUCTION

Recent advances in wireless communication have enabled distributed sensing and actuation through wireless sensor and actuator networks (WSANs) [1]. However, these networks, which consist of wirelessly connected sensor and actuator nodes, are constrained by communication range, energy, and infrastructure costs. Conventional nodes with limited battery life need frequent replacements, which is infeasible in hazardous deployments. To overcome these issues, emerging technologies such as backscatter communication (BSC) and wireless energy transfer (WET) are of interest [2], [3].

In WSANs, a key objective is to send sensor data to actuators for timely action while ensuring the availability of required energy. The Age of Information (AoI) approach focuses only on optimizing sensor updates [4]. In [5], AoI is combined with data uncertainty; but AoI alone does not ensure the timeliness of data-driven action. In [6], the actuation delay is modeled by assuming direct communication between the sensor and actuator, which may not be valid in remote deployments. The work in [7] proposed using a controller for data relaying. However, a static controller may not be sufficient for communication and actuation operations, as these power-intensive tasks require the controller to be close to the sensor and actuator nodes. To this end, unmanned autonomous vehicles (UAVs) can be used in BSC and WET based WSAN.

Although UAV-based aerial communication has been widely studied, UAV-aided sensor data relaying and WET has not been explored before. UAV offers operational flexibility and

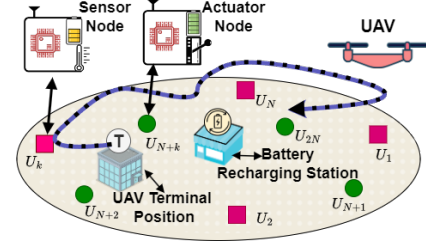


Figure 1 UAV and BSS aided wireless sensor actuator network.

sustainability in WSAN through WET and monostatic BSC, but it is constrained by its limited onboard battery and energy losses due to imprecise UAV localization and low WET efficiency. While UAV energy limitation can be addressed by placing a battery swap station (BSS) within the region [8], [9], optimizing UAV-aided WSAN operation considering the various system constraints requires in-depth analysis.

Building upon the challenges outlined above, the key contributions of this work are as follows: (1) Using tools from stochastic geometry, closed-form expressions are derived for the BSC ergodic capacity \bar{r}_{BSC} and the expected energy harvesting rate \bar{P}_{EH} , incorporating Nakagami- m channel fading and UAV location uncertainty. (2) Based on \bar{r}_{BSC} and \bar{P}_{EH} and WSAN requirements a novel UAV-assisted framework is proposed to minimize the actuation delay in WSAN. (3) To achieve this objective, the problem is reformulated into an equivalent form that optimizes the UAV node visiting sequence. (4) Given the high complexity induced by the constrained combinatorial nature of the problem, a sequential deep reinforcement learning (SDRL) approach is employed for solving it. (5) Finally, the proposed scheme is shown to offer reduced actuation delay at a significantly smaller computational overhead compared to the benchmark methods.

We remark that this kind of mobile-aided autonomous WSAN system can be useful in various practical situations. For instance, sensor data from avalanche or landslide-prone areas can be used to actuate early warning signs nearby the affected road sections, for enhancing both safety and convenience.

II. SYSTEM MODEL

Consider a UAV-aided WSAN, having N sensor-actuator pairs, a UAV, and a BSS, deployed in a circular region of radius R_{max} (Fig. 1). The nodes are assumed to be located far apart; thus, the UAV can serve only one node at a time. The UAV is required to visit each sensor node before the corresponding actuator, but only once in each round. However, it may visit the BSS as often as needed. The UAV flies at a constant speed V_{UAV} and an altitude h_{UAV} . Its power consumptions for movement P_{mov} and hovering P_{hov} are modeled as in [10].

The sensor and actuator nodes have both communication and energy harvesting modules. Each sensor node also contains a sensing module, whereas each actuator node has a mechanical actuation module. The combined set of sensor and actuator devices is denoted as $\mathcal{U} = \{1, \dots, 2N\}$. The index $k \in \{1, \dots, N\}$ denotes the k -th sensor, and the index $k + N \in \{N + 1, \dots, 2N\}$ denotes the actuator associated with the k -th sensor. The nodes are static, and their locations are known apriori. The coordinates of n -th node is denoted as $w_n = \{x_n, y_n, z_n\}$. For convenience, indices '0' and '2N + 1' represent the BSS and the UAV terminal station with respective coordinates $\{x_0, y_0, z_0\}$ and, $\{x_{2N+1}, y_{2N+1}, z_{2N+1}\}$. The 2D Euclidean distance between any two indices is denoted as d_{ij} .

The UAV collects the sensor data using BSC by hovering over it and transmitting an unmodulated signal x_1 toward the sensor node. The received signal at the sensor node is $y_S = h_1 \sqrt{P_{tx}} x_1 + n_1$. The sensor node reflects back the received signal after modulating it with the information signal x_2 . The signal received by the UAV is [11]: $y_U = h_2 h_1 \sqrt{P_{tx}} x_1 x_2 + n_1 + n_2$, where P_{tx} is the transmit power of the UAV, n_1 and n_2 denote additive white gaussian noise (AWGN). There can be some imperfection in hovering position due to UAV localization error, which are discussed in Section III. The wireless communication link is primarily line-of-sight, and the channel coefficients follow a Nakagami- m distribution. The channel gain is $h_i = \sqrt{G_1 G_2 (\lambda/4\pi)^2 d^{-\alpha_p} \tilde{h}_i} \forall i \in \{1, 2\}$. Here, \tilde{h}_i denotes the small-scale fading coefficients, α_p is the path loss exponent, λ is the transmission frequency, and G_1, G_2 respectively denote the UAV and node antenna gain. The communication data rates from the sensor node to UAV r_{S-U} and the UAV to actuator node r_{U-A} are expressed as

$$r_{S-U} = \log_2 \left(1 + \frac{P_{tx} |h_1|^2 |h_2|^2}{\sigma_1^2 + \sigma_2^2} \right), r_{U-A} = \log_2 \left(1 + \frac{P_{tx} |h_1|^2}{\sigma_1^2} \right). \quad (1)$$

The harvested power P_{EH} , with $P_{rx} = P_{tx} |h_1|^2$, is modeled as [12]

$$P_{EH}(P_{rx}) = (a P_{rx} + b) \cdot \mathbf{1}_{[\varrho_1, \varrho_2)}(P_{rx}) + P_{EH, \max} \cdot \mathbf{1}_{[\varrho_2, \infty)}(P_{rx}) \quad (2)$$

where $P_{EH, \max}$ is the maximum possible harvest power, ϱ_1 is the receiver sensitivity, while a and b are shaping parameters.

III. STATISTICAL ANALYSIS

In this section, \bar{r}_{BSC} and \bar{P}_{EH} are derived based on the above channel model and UAV location uncertainty.

Distance distribution: While serving the n -th node, the UAV is meant to hover above w_n . Due to location error, its actual position is considered uniformly distributed in a ball $B(w_n, R_{\max})$. Thus, the UAV-to-node distance distribution is

$$f_d(d) = \frac{2d}{\sqrt{d^2 + h_{UAV}^2}}; \quad h_{UAV} \leq d \leq \sqrt{h_{UAV}^2 + R_{\max}^2} \quad (3)$$

BSC ergodic capacity: It is defined as maximum achievable average data rate, taking into account both the random variations in the wireless fading channel, which is modeled as a gamma-distributed random variable, and the spatial distribution of the UAV-to-node distance as modeled in (3) and is expressed as $E_d [E_{h_1 h_2} [\log_2 (1 + \gamma)]]$.

Theorem 1. *The closed-form expression for BSC ergodic capacity is expressed in (4), where*

$$C_4 = \frac{\alpha_1 \alpha_2 P_{tx} (G_1 G_2)^2 (\lambda/4\pi)^4}{m_1 m_2}, C_3 = \frac{1}{\Gamma m_1 \Gamma m_2}.$$

Proof. See Appendix A. ■

Expected energy harvesting rate: It is defined as the average amount of power that can be harvested by the node via WET process.

Theorem 2. *The closed-form expression considering Gamma-distributed wireless fading link gain and spatial distribution as modeled in (3) is expressed in (5), where $K = P_{tx} E[d^{\alpha_p}] (\lambda/4\pi)^2 G_1$.*

Proof. See Appendix B. ■

IV. PROBLEM FORMULATION

We now formulate the actuation delay minimization based on the node requirements and statistical measures from (4) and (5). Then, the associated feasibility constraints are presented.

The actuation delay minimization is formulated as a sequence prediction problem. After the UAV leaves the terminal, it travels to each sensor node to gather data and proceeds to the associated actuator node to transfer information and energy. This process continues until data are collected from all sensor nodes and the respective actuators are served, and the UAV recharges itself as required. UAV visiting sequence is denoted as $S = \{s_k\}_{k=1}^L$ where $s_k \in \{0, 1, \dots, 2N + 1\}$ denotes the indices having one-to-one mapping with the devices, BSS, and terminal station, L is the sequence length, which is at least $2(N + 1)$, but could be longer since the UAV may need to visit the BSS multiple times. S needs to be optimized for reduced actuation delay. In finding the optimal S , the feasibility needs to be ensured, as the constraints can affect the viability of S .

A. Feasible sequence

We define a variable $\mathbf{I}(k)$ that indicates the index of the location visited by the UAV at k -th position in the sequence and is defined as $\mathbf{I}(k) = m$, s.t. $s_m = k$; $m \in \{1, 2, \dots, L\}$, $k \in \{1, 2, \dots, 2N\}$. While serving, each device should be visited only once, with an actuator visited only after its corresponding sensor is served. The constraints are represented as

$$\text{if } s_i = k \Rightarrow s_j \neq k; \forall j \neq i, \forall i, j \in \{1, \dots, L\}, k \in \{1, \dots, 2N\} \quad (6)$$

$$\mathbf{I}(k) < \mathbf{I}(N + k) \quad \forall k \in \{1, \dots, N\} \quad (7)$$

Moreover, the total time the UAV spends on each device is required to be greater than or equal to the required service time for that device. The variable τ_j denotes the expected time required to serve the j -th device and is defined as

$$\tau_j = \begin{cases} \frac{D_j^{\text{req}}}{\bar{r}_{S-U,j}} \left(1 + \frac{P_c}{P_{EH}}\right) & \forall j \in \{1, \dots, N\} \\ \frac{E_j^{\text{req}}}{P_{EH}} + \frac{D_j^{\text{req}}}{\bar{r}_{U-A,j}} \approx \frac{E_j^{\text{req}}}{P_{EH}} & \forall j \in \{N + 1, \dots, 2N\} \\ \tau_{BSS} & j = 0 \end{cases} \quad (8)$$

where D_j^{req} denotes the amount of data required to be transmitted, E_j^{req} denotes the required energy for actuation, P_c denotes

$$\bar{r}_{\text{BSC}} = \frac{C_3 h_{\text{UAV}}^2 R^3}{2} H_{1,0,4,3;1,2}^{0,1,1,4;1,1} \left(\begin{matrix} (0, 2, 1) \\ - \end{matrix} \middle| \begin{matrix} (1 - m_1, 1) (1 - m_2, 1) (1, 1) (1, 1) \\ (1, 1) (1/2, 2) (0, 1) \end{matrix} \right) \left(\begin{matrix} (-1/2, 1) \\ (-1, 1) (-3/2, 1) \end{matrix} \right) C_4 h_{\text{UAV}}^{-4} \frac{R^2}{h_{\text{UAV}}^2} \quad (4)$$

$$\bar{P}_{\text{EH}} = \frac{a\alpha_1 K}{m_1 \Gamma m_1} \left(\gamma \left(m_1 + 1, \frac{m_1 \varrho_2}{K \alpha_1} \right) - \gamma \left(m_1 + 1, \frac{m_1 \varrho_1}{K \alpha_1} \right) \right) + \frac{b}{\Gamma m_1} \left(\gamma \left(m_1, \frac{m_1 \varrho_2}{K \alpha_1} \right) - \gamma \left(m_1, \frac{m_1 \varrho_1}{K \alpha_1} \right) \right) + \frac{P_{\text{EH}, \max}}{\Gamma m_1} \left(1 - \frac{1}{\Gamma m_1} \gamma \left(m_1, \frac{m_1 \varrho_2}{K \alpha_1} \right) \right) \quad (5)$$

the sensor node circuit power consumption while performing BSC, and τ_{BSS} denotes the time required for battery swapping. Thereby, total energy required to spend by UAV on j -th device is expressed as $\mathcal{E}_j^{\text{ser}} \approx (P_{\text{Hov}} + P_{\text{tx}}) \tau_j \quad \forall j \in \{1, \dots, N\}$. The UAV transmits with P_{tx}^a for actuator nodes and P_{tx}^s for sensor nodes. Once the node has been served, $\mathcal{E}_j^{\text{ser}}$ is reset to null. Furthermore, throughout the process, it is necessary to monitor the on-board battery and recharge it when required. The variable $B_{\text{rem},j}$ denotes the remaining on-board battery capacity of UAV after it has completed traversing the sequence upto j -th element $\forall j \in \{0, 1, \dots, 2N + 1\}$. Furthermore, to represent the relationship between two nodes that are visited in sequence, a binary variable $\mathcal{X}_{ij}(n)$ is defined as $\mathcal{X}_{ij}(n) = 1$ if $s_n = i$ and $s_{n+1} = j$, where, $i, j \in \{0, \dots, 2N + 1\}$ and $n \in \{0, \dots, L - 1\}$. Given that $B_{\text{rem},n}$ is known and $\mathcal{X}_{ij}(n) = 1$ where $n \in \{0, \dots, L - 1\}$, then the battery status is updated as

$$B_{\text{rem},n+1} = \begin{cases} B_{\text{rem},n} - \mathcal{E}_{ij}^{\text{mov}} - \mathcal{E}_j^{\text{ser}} & j \in \{1, \dots, 2N\} \\ B_{\text{full}} & j = 0 \\ B_{\text{rem},n} - \mathcal{E}_{ij}^{\text{mov}} & j = 2N + 1 \end{cases} \quad (9)$$

It is to be noted that the onboard battery status restricts the UAV movement. The constraints are represented as

$$\mathcal{X}_{ij}(n) \geq 0 \text{ if } \begin{cases} B_{\text{rem},n} \geq \mathcal{E}_{ij}^{\text{mov}} + \mathcal{E}_j^{\text{ser}} + \mathcal{E}_{j0}^{\text{mov}} & j \in \{1, \dots, 2N\} \\ B_{\text{rem},n} \geq \mathcal{E}_{ij}^{\text{mov}} & j \in \{0, 2N + 1\} \end{cases} \quad (10)$$

$$\mathcal{X}_{i0}(n) = 1 \text{ if } B_{\text{rem},n} - \mathcal{E}_{i,0} < \min \{ \mathcal{E}_{ij}^{\text{mov}} \} \quad \forall n \in \{1, \dots, L - 1\}, \quad (11)$$

$$i \in \{1, \dots, 2N\}, j \in \{1, \dots, 2N + 1\} \setminus \{i\}$$

where, (10) indicates that after serving the i -th node the UAV can visit the j -th node only if the remaining battery capacity is sufficient for both serving the j -th node and returning to the BSS from j -th node. Moreover, (11) states that if the remaining battery capacity is not enough to serve any unserved devices, the UAV should proceed to the BSS. The subtraction in (11) ensures that the UAV always has sufficient battery to reach BSS. Furthermore, (12) highlights that the UAV in any feasible sequence starts and ends on UAV terminal position.

$$\sum_i \mathcal{X}_{2N+1,i}(0) = 1, \sum_i \mathcal{X}_{i,2N+1}(L - 1) = 1 \quad \forall i \in \{0, \dots, 2N\} \quad (12)$$

B. Problem formulation

The UAV performance is measured by the maximum delay of actuation (MDA), defined as the maximum time difference between the start of operation and when the actuator takes action. The overall optimization problem considering this metric and the feasibility constraints is expressed as

$$(\mathcal{P}1): \quad \min_S \quad \max_k |t_{k+N} - t_0| \quad (13)$$

$$\text{s.t. :} \quad (4), (7), (10) - (12)$$

where, t_0 is the time at which UAV starts the operations and t_{k+N} denotes the time at which actuator associated

with k -th sensor takes action which is expressed as $t_j = \sum_{i=1}^{\mathbf{I}(j)} (d_{s_i, s_{i+1}} / V_{\text{UAV}}) + \tau_i, \forall j \in \{N + 1, \dots, 2N\}$. The problem $\mathcal{P}1$ is an NP-hard combinatorial optimization problem. Considering the UAV battery capacity sufficiently large and homogeneity among all the nodes, the transformed problem becomes an equivalent NP-hard problem as in [10]. Consequently, the problem $\mathcal{P}1$ is inherently NP-hard, rendering it difficult to address through conventional solution methods. Therefore, we propose a sequential deep reinforcement learning approach building on [13], [14] to effectively address this challenge. The details of the solution are outlined in the following section.

V. PROPOSED SDRL-BASED SOLUTION

The proposed solution utilizes a Markov Decision Process framework, where the UAV (agent) interacts with the dynamic environment by selecting actions that generate rewards while transitioning among states. Here, the UAV is responsible for making decisions and executing actions. A state S_l includes information about the last visited node and relevant parameters such as $\{w_n, \tau_n, \mathcal{E}_n^{\text{ser}}\}_{n=0}^{n=2N+1}$, and $B_{\text{rem},l}$. The action set consists of the available choices at a given state S_l , which includes unvisited node indices, BSS, and the UAV terminal station, all subject to the constraints defined in $\mathcal{P}1$. The action a_l represents the index of the location the UAV visits. The reward is calculated at the end of the episode and is defined as the negative of the maximum actuation delay, as described in $\mathcal{P}1$.

A. Sequential neural network architecture

To map the current state to a probability distribution of possible actions, a sequential neural network-based policy is used at the UAV. Primarily, sequential neural network architecture is composed of an encoder and decoder module.

Encoder: The encoder input consists of static components $C_{s,n} = \{w_n \cup \tau_n\}$ (location and expected serving time) and dynamic components $C_{d,n} = \{B_{\text{UAV}} \cup \mathcal{E}_n^{\text{ser}}\} \quad \forall n = \{0 \dots 2N + 1\}$ (UAV battery and energy requirements). These are processed through an embedding layer comprising a convolutional encoder, which maps the low-dimensional data to a high-dimensional space and produces embedded outputs C_{emb,s_n} and C_{emb,d_n} , with total inputs and outputs denoted as C_{in} and C_{emb} .

Decoder: For a policy ϕ , the probability that the UAV follows a sequence S conditioned on C_{emb} is defined as $P_\phi(S|C_{\text{emb}}) = \prod_{l=1}^L P(s_{l+1}|S_l, C_{\text{emb}})$, where S_l denotes the sequence up to l steps. At each step $l \in \{1, \dots, L\}$, the decoder generates the conditional probability distribution $P(s_{l+1}|S_l, C_{\text{emb}})$, that determines the agent action a_l . The optimal policy ϕ^* gives the optimal sequence S^* with probability 1. We aim to minimize the optimality gap between ϕ and ϕ^* .

In the l -th decoding step, the static information of the last visited index is encoded as $C_{\text{emb},s_{l-1}}$, and Gated Recurrent Unit (GRU) processes this encoded data along with the hidden state

Algorithm 1 Pseudo-code for SDRL training stage

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1: Input: batch size  $\mathcal{B}$ , training dataset  $\mathcal{D}$ .
2: Output: Policy parameters  $\varrho_a$ ,  $\varrho_c$ 
3: for  $\text{itr} = 1, 2, \dots, \text{itr}_{\text{tot}}$  do
4:   reset gradient  $d\varrho_a \leftarrow 0, d\varrho_c \leftarrow 0$ 
5:   Obtain training batch  $\{\mathcal{D}_{(\text{itr} \times \mathcal{B})+1}, \dots, \mathcal{D}_{(\text{itr}+1) \times \mathcal{B}}\}$ 
6:   for  $b = 1, 2, \dots, \mathcal{B}$  do
7:      $l \leftarrow 0$ 
8:     do
9:       compute  $P(s_{l+1}^b | S_l^b, C_{emb}^b)$ 
10:      choose action accordingly
11:       $l \leftarrow l + 1$ 
12:      while terminal state is reached
13:      compute reward  $R^b$ 
14:    end for
15:    calculate gradient
16:     $d\varrho_a \leftarrow \frac{1}{\mathcal{B}} \sum_{b=1}^{\mathcal{B}} (R^b - V^b) \nabla_{\varrho_a} \log(P(s | C_{emb}))$ 
17:     $d\varrho_c \leftarrow \frac{1}{\mathcal{B}} \sum_{b=1}^{\mathcal{B}} \nabla_{\varrho_c} (R^b - V^b)^2$ 
18:    Update  $\varrho_a$  and  $\varrho_c$ 
19:  end for

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h_{l-1} maintained by GRU, yielding $h_l, Y_l = \text{GRU}(h_{l-1}, C_{emb, s_{l-1}})$. The GRU output Y_l is then passed through the attention block which incorporates the importance of the locations at the current decoding step. This is achieved through the following operation: $a_l = \text{softmax}(v_{a1}^T \tanh(W_{a1} \cdot C_{emb}) + v_{a2}^T \tanh(W_{a2} \cdot Y_l))$. Using a_l , a context vector c_l , is generated as $c_l = \sum_{i=0}^{2N+1} (a_l^i C_{emb, i})$. The encoded output C_{emb} is then combined with c_l to produce κ_l as $\kappa_l = v_{c1}^T \tanh(W_{c1} \cdot C_{emb}) + v_{c2}^T \tanh(W_{c2} \cdot c_l)$. Before generating the action probability distribution, the masking vector \mathcal{M}_l is applied to κ_l as $\kappa_l = \kappa_l + \mathcal{M}_l$, where $\mathcal{M}_{l,i} \in \{0, -\infty\} \forall i \in \{1, \dots, 2N+2\}$, enforces that the feasibility conditions in $\mathcal{P}1$, are satisfied. In addition, the actuator nodes are masked until all the sensor nodes are visited. Finally, the conditional probability distribution which determines the agent action a_l , is calculated as $P(s_{l+1} | S_l, C_{emb}) = \text{softmax}(\kappa_l)$. The process repeats until the termination condition is satisfied.

B. Training

The UAV policy network is trained using the REINFORCE algorithm to maximize the expected cumulative reward, as outlined in Algorithm 1. The algorithm utilizes both an actor and a critic neural network to optimize the policy. The actor adjusts the policy parameters ϱ_a , increasing the likelihood of selecting actions that lead to positive outcomes. During each episode, the agent collects experience from the environment and computes the cumulative reward R^b at the end, which serves as a performance estimate. Following the policy gradient approach, the policy parameters are then updated in the direction that increases the likelihood of actions that lead to higher rewards. The critic estimates the value function V^b and associated parameters ϱ_c are trained using stochastic gradient descent on the mean squared error (MSE) between predicted value and actual reward, helping to reduce variance in policy updates. To enhance learning efficiency, parallel learning and batch processing are used, addressing correlation between successive states and improving training stability.

Table I Simulation parameters and settings [9], [12], [13]

Parameter	Value	Parameter	Value
m_1, m_2	2	D_{req}	24-40 Bytes
α_1, α_2	1	E_{req}	31-47 mJ
G_1	10 dBi	τ_{BSS}	180 s
G_2	0 dBi	h_{UAV}	10 m
σ_1^2, σ_2^2	1×10^{-9}	a	0.24714,
λ	915 MHz	b	-1.5817e-5
R_{max}	150 m	PEH_{max}	0.00492 μ W
ΔR	2 m	P_c	10.6 μ W
α_p	2	ϱ_1, ϱ_2	0.000064, 0.02
P_{tx}^a	40 W	V_{UAV}	9.8 m/s
P_{tx}^s	10 W	Training size	128e+4
Validation size	1e+3	Convolution layers	1
Learning rate	5e-4	Hidden size h_s	128
Batch size	256	Dropout rate	0.1

Table II Compute overhead for different sensor-actuator pair counts

Method	Computation Time (ms)		
	N = 10	N = 15	N = 20
Baseline	0.1675	0.19	0.26
HSR	1504	2100	3080
SDRL	86.76	104.96	134.65

VI. NUMERICAL RESULTS

The SDRL simulations are performed in Python with the PyTorch library on a system having an Intel Xeon W-2145 3.70 GHz processor, 64 GB RAM, and Nvidia Quadro P5000 GPU. The values of the simulation set-up parameters are listed in Table I. Fig. 2(a) highlights the BSC ergodic capacity. The analytical results align with the Monte Carlo simulations, highlighting the accuracy of the derived analytical expressions. The plot illustrates a positive correlation between ergodic capacity and transmit power while also showing a reduction with increasing hovering altitude due to higher path loss.

Fig. 2(b) illustrates the performance of the UAV energy transfer capability in terms of expected energy harvesting rate. The plot highlights the effects of UAV altitude variation and localization inaccuracy. At lower elevations, the impact of localization uncertainty on performance is more pronounced. Fig. 2(c) shows the learning curves of the proposed algorithm for different battery capacities. The curves demonstrate eventual convergence to the long-term return. The maximum delay is observed with the lowest battery capacity.

Figs. 2(d) and 2(e) compares the proposed algorithm, the baseline solution, and the hybrid swap-and-reverse (HSR) algorithm. The baseline solution satisfies the feasibility constraints but does not necessarily yield an optimal result. In contrast, the HSR algorithm is based on 2-Opt operation [15], begins with a feasible solution, and randomly selects two indices. Based on a specified probability, either the nodes at these indices are swapped, or the sequence between them is reversed. The resulting sequence is retained if it yields higher reward and remains feasible; otherwise, it is discarded. This iterative process continues to progressively refine the solution.

Fig. 2(d) demonstrates that, for a fixed battery size of 80 kJ, the maximum actuation delay increases as the number of sensor-actuator node pairs rises. In Fig. 2(e), where simulations are performed with 20 sensor-actuator pairs for varying battery capacities, a clear reduction in maximum actuation delay is observed as battery capacity increases. In both cases, the proposed algorithm outperforms the baseline and HSR

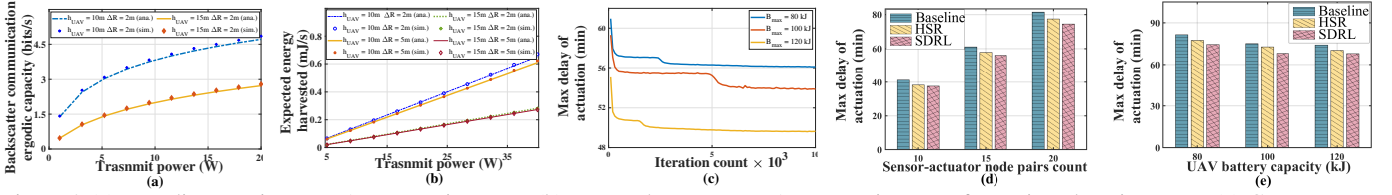


Figure 2 (a) Ergodic capacity vs. UAV transmit power. (b) Energy harvest vs. UAV transmit power for various location error. (c) Convergence of proposed SDRL algorithm; $N = 15$. (d) Actuation delay, with battery size 80 kJ. (e) Actuation delay for 20 sensor-actuator node pairs.

algorithms, improving performance and significantly reducing computation time compared to HSR. This time reduction is summarized in Table II, highlighting the efficiency of the proposed approach. The proposed SDRL order complexity is $O(LN h_s^2 C_1)$ while the complexity of HSR is $O(I_{max} N^2 C_2)$, where C_1, C_2 are constants, h_s is the hidden size, and I_{max} is the number of iterations. Furthermore, the proposed algorithm is robust to variations in the number of sensor-actuator node pairs, meaning it does not require retraining when there are changes in the node count or the node configuration. This characteristic enhances its generalizability, allowing it to effectively adapt to modifications in network requirements.

VII. CONCLUSION

This paper introduced UAV and BSS-aided WSN for minimizing actuation delay with imperfect UAV localization and wireless fading channel constraints. Closed form expressions of BSC ergodic capacity for based data collection and expected energy harvesting rate were derived. Based on node requirements and the statistical properties of communication and energy harvesting, the node visit sequence of UAV was optimized using sequential DRL. The simulation results demonstrated that the proposed strategy offers reduced actuation delay with a significantly lower computation overhead.

APPENDIX

A. Proof of (4)

The pdf of the product of two independent but not identical (i.n.i.d.) gamma random variables h_1 and h_2 is

$$f_{h_1 h_2}(z) = \frac{2}{\Gamma m_1 \Gamma m_2} \left(\frac{m_1 m_2}{\alpha_1 \alpha_2} \right)^{\frac{m_1 + m_2}{2}} z^{\frac{m_1 + m_2 - 2}{2}} K_{m_1 - m_2} \left(2\sqrt{\frac{m_1 m_2 z}{\alpha_1 \alpha_2}} \right). \quad (\text{A.1})$$

The ergodic capacity is defined as

$$\frac{1}{\log 2} \int_{d_{\min}}^{d_{\max}} \int_0^{\infty} \log(1 + \gamma) f_{h_1 h_2}(x) f_d(d) dx dd \quad (\text{A.2})$$

where $\gamma = \eta d^{-2\alpha} x$, using (A.1) and [16, 07.34.03.0456.01]. Consider the following integral:

$$I_1 = C_1 \int_0^{\infty} G_{2,2}^{1,2} \left(\frac{11}{10} \middle| C_2 x \right) x^{\frac{m_1 + m_2 - 2}{2}} K_{m_1 - m_2} (2\sqrt{x}) dx. \quad (\text{A.3})$$

Using [17, Eq. (7.821.3)], and substituting in (A.2) we have

$$\int_{d_{\min}}^{d_{\max}} C_3 G_{4,2}^{1,4} \left(\frac{1 - m_1, 1 - m_2, 1, 1}{1, 0} \middle| C_4 d^{-2\alpha} \right) \frac{d}{\sqrt{d^2 - h_{\text{UAV}}^2}} dd \quad (\text{A.4})$$

where $d_{\min} = h_{\text{UAV}}$, $d_{\max} = \sqrt{R^2 + h_{\text{UAV}}^2}$. Thereafter, using [17, Eq. (9.34.7)] we get (4).

B. Proof of (5)

The expected energy harvesting rate is

$$\bar{P}_{\text{EH}} = \int_{\mathcal{Q}_1}^{\mathcal{Q}_2} (aP + b) f_{\bar{P}_{\text{rx}}}(P) dP + P_{\text{EH}, \max} \int_{\mathcal{Q}_2}^{\infty} f_{\bar{P}_{\text{rx}}}(P) dP \quad (\text{B.1})$$

where $\bar{P}_{\text{rx}} = E_d[P_{\text{rx}}] = G_1 (\lambda/4\pi)^2 E[d^{-\alpha P}] h_1 P_{\text{tx}}$. For $\alpha P = 2$, after simple mathematical calculations, we get

$$E[d^{-\alpha P}] = \int_{d_{\min}}^{d_{\max}} d^{-\alpha P} f_d(d) dd = \frac{2}{h_{\text{UAV}} R_{\max}} \tan^{-1} \left(\frac{R_{\max}}{h_{\text{UAV}}} \right). \quad (\text{B.2})$$

Finally, substituting (B.2) into (B.1) and using [17, Eq. (8.350.1)], the close form expression (5) is obtained.

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